

# Phase-Lag Compensator

Example  $G(s) = \frac{100}{s(s+4)(s+5)}$

Task: Design a compensator

Show the Bode plots with  $PM = 18^\circ$ :

margin (100, [1, 9, 20, 0])

The PM of at least  $40^\circ$  is considered adequate

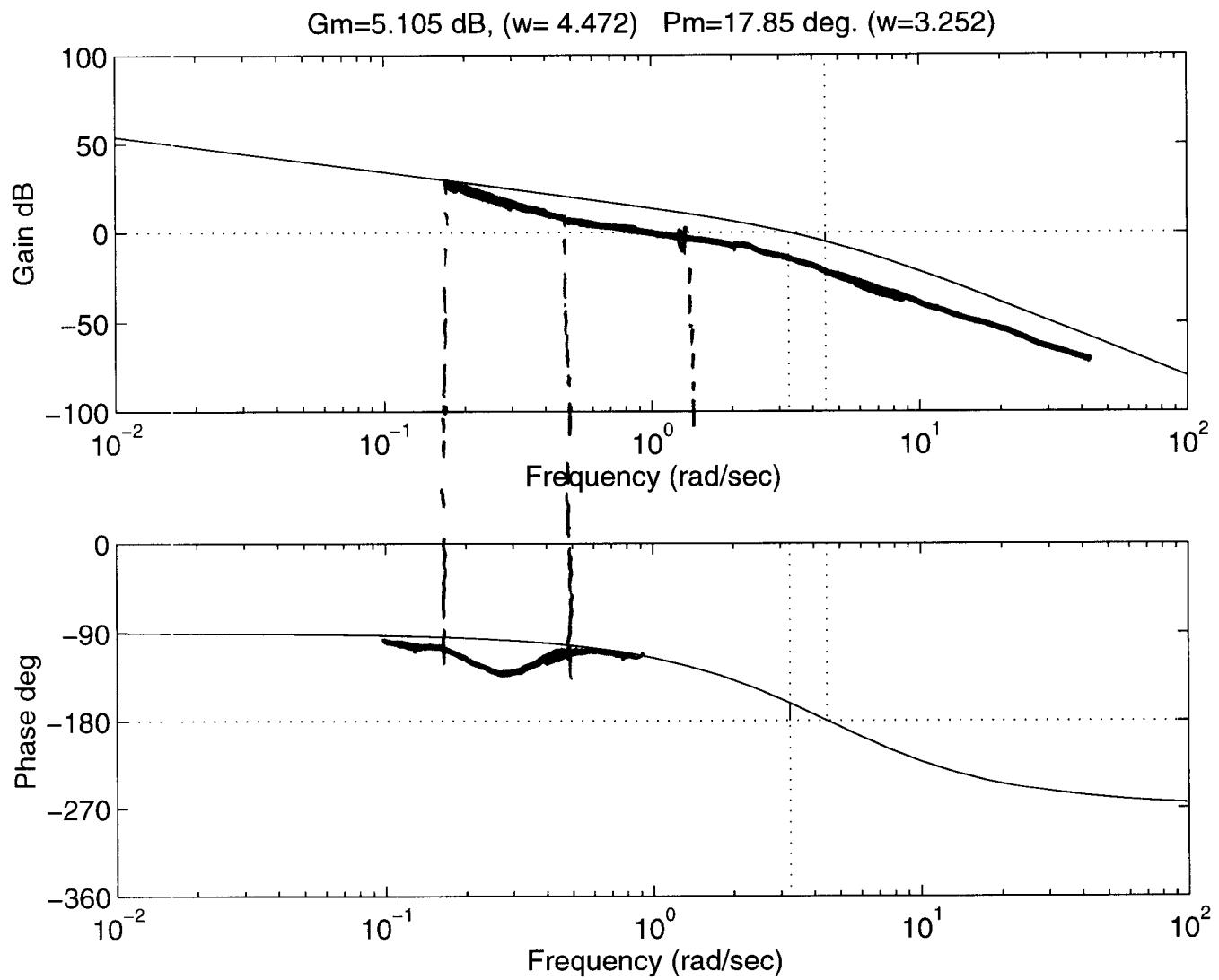
Task: design a compensator which increases the phase margin.

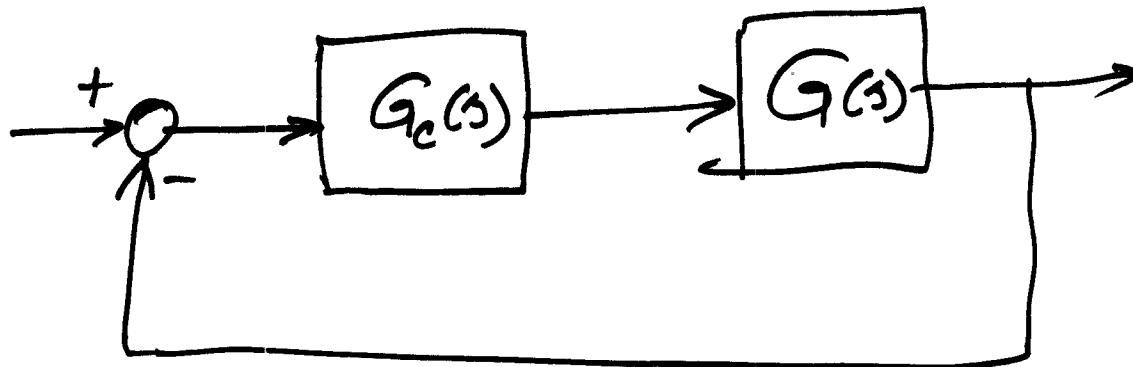
Observation: We want to reduce the amplitude response without affecting much the phase response (at the frequencies of interest)

Solution: ~~a pole~~ pole followed by a zero at low freq.

7a

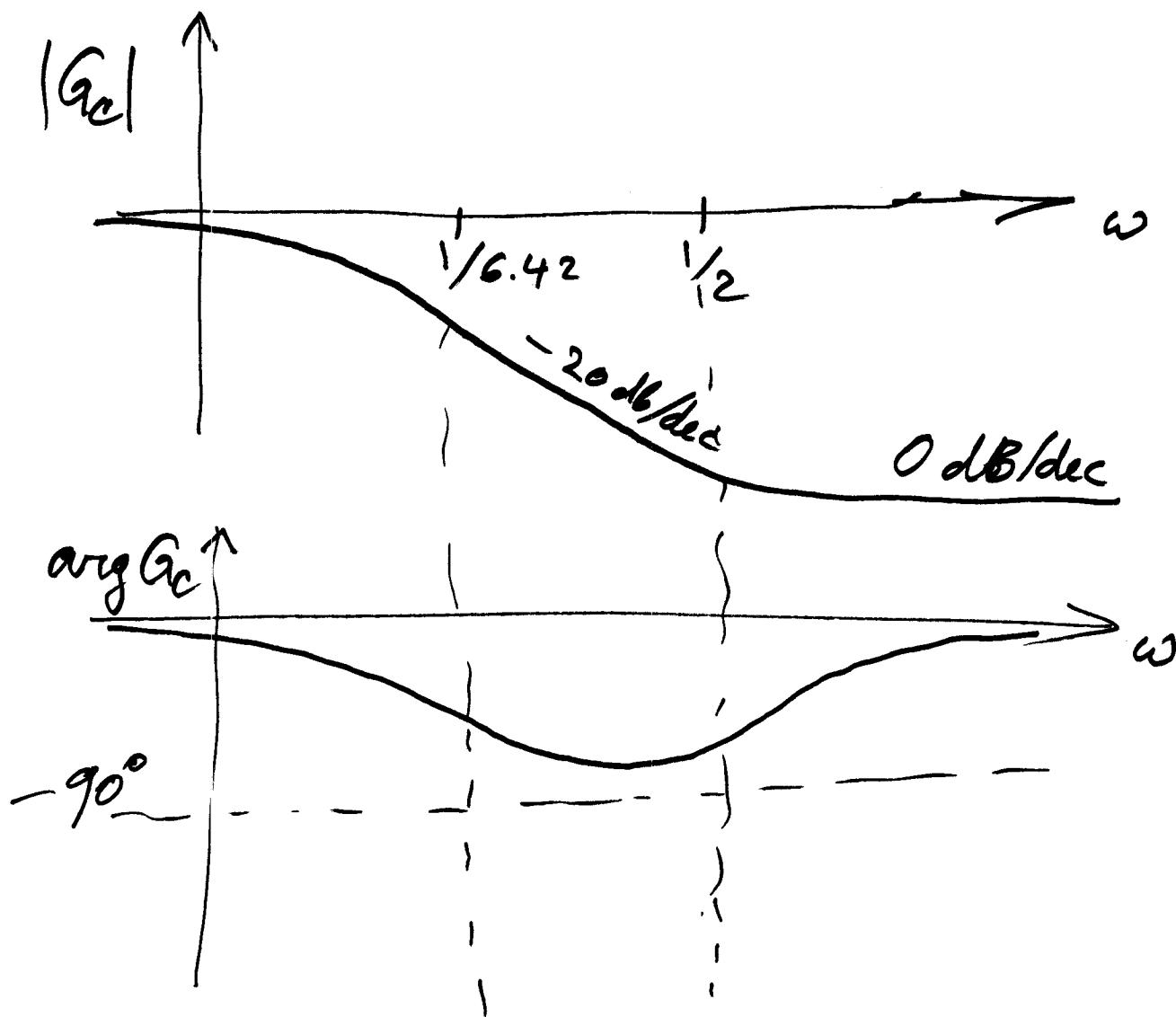
$$G(\beta) = \frac{100}{\beta(\beta+4)(\beta+5)}$$





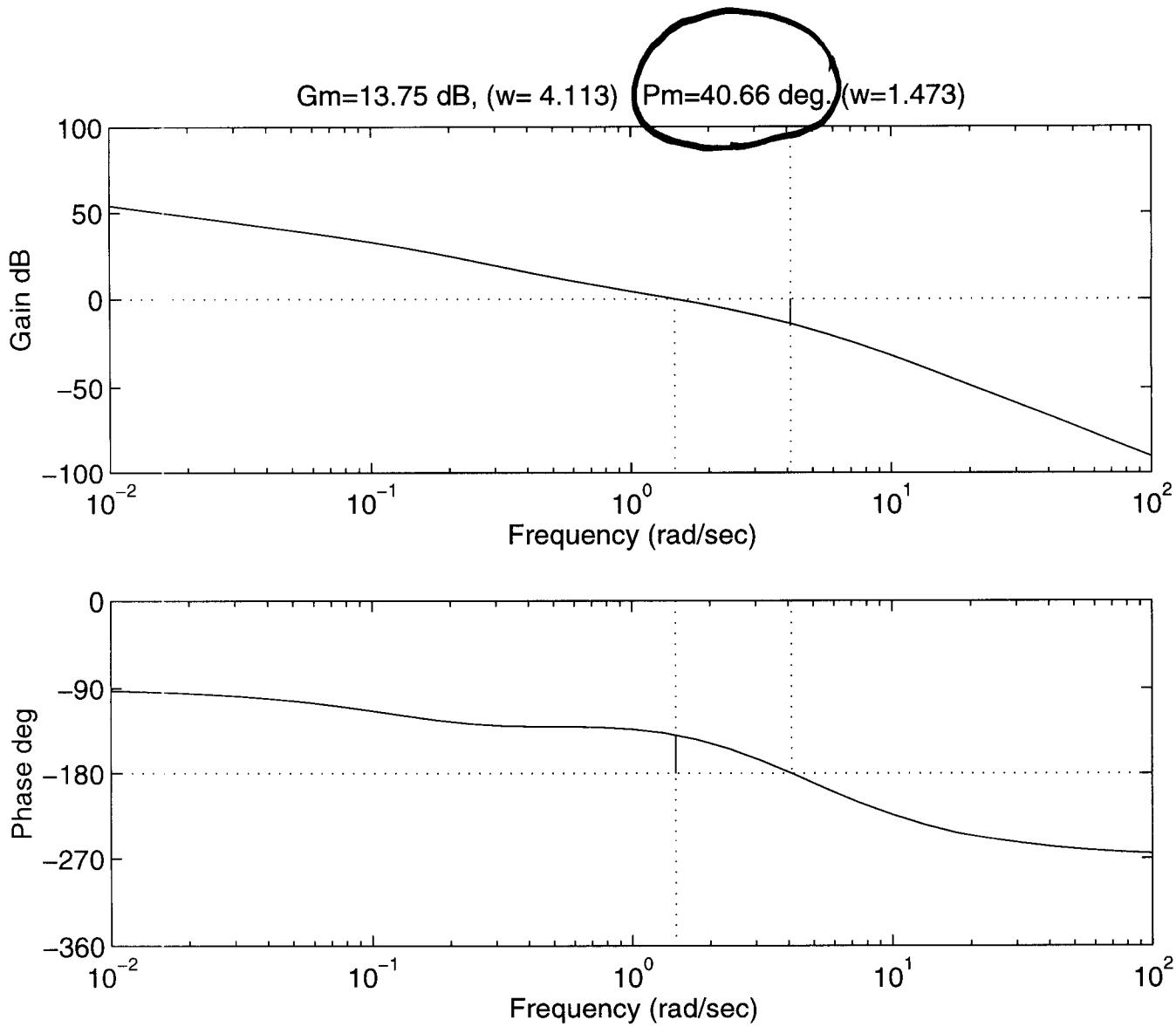
We pick

$$G_c(s) = \frac{2s+1}{6.42s+1} \quad \underline{\text{LAG}}$$



8a

$$G_c(s) G(s) = \frac{25+1}{6.425+1} \frac{100}{s(s+4)(s+5)}$$



$$[n, d] = \text{series}([2 1], [6.42 1], 100, [1 9 20 0])$$

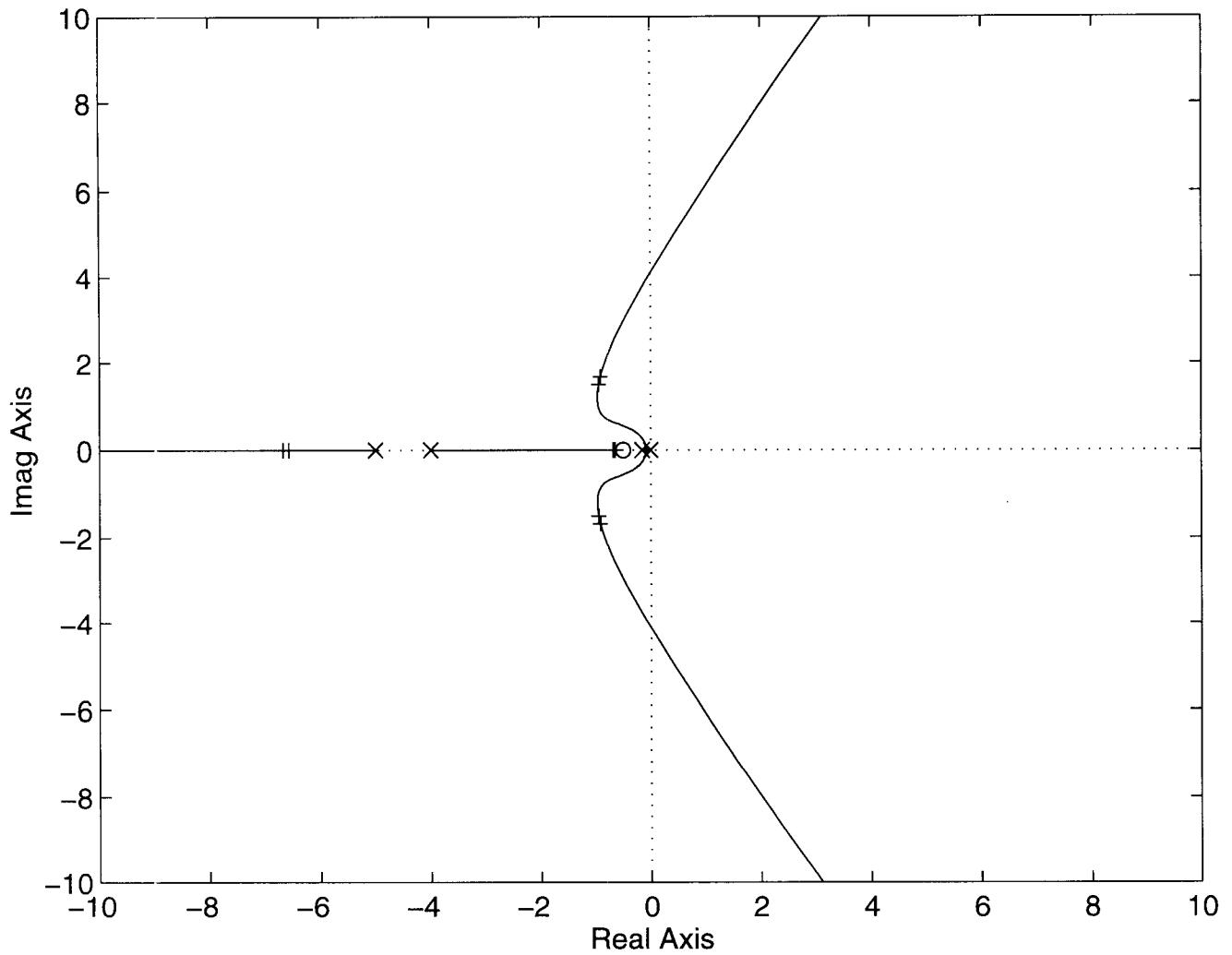
margin (n, d)

In general, phase-lag compensators have the form

$$G_c(s) = \frac{\alpha Ts + 1}{Ts + 1}, \quad \alpha < 1$$

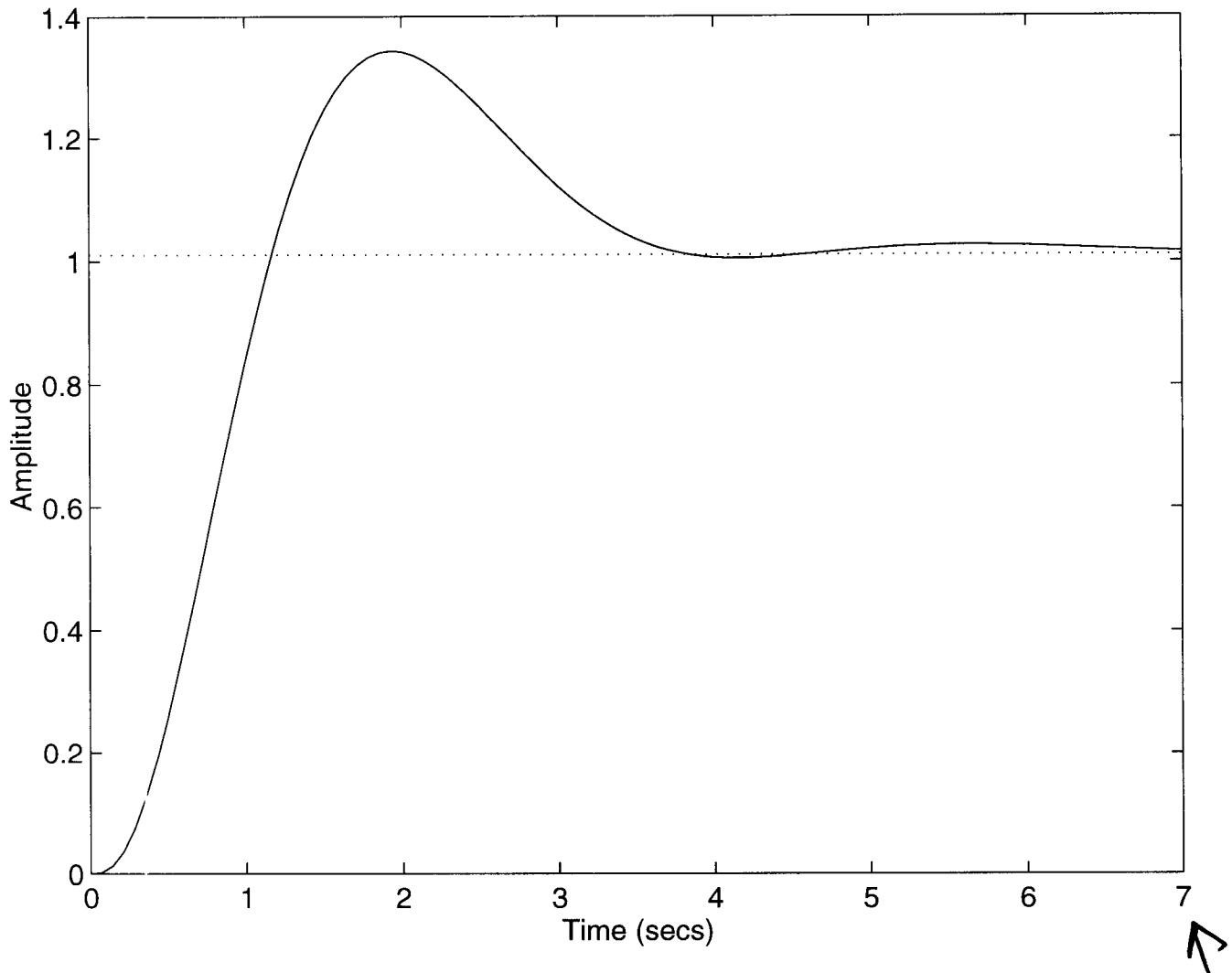
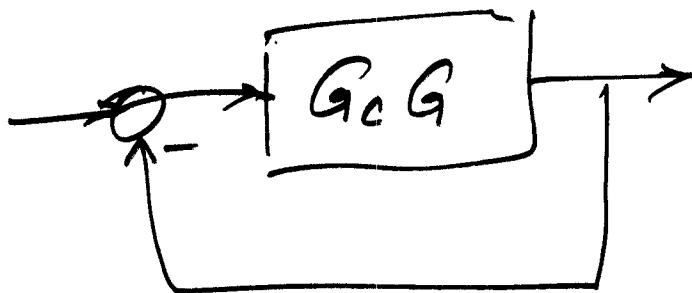
Downside of phase-lag compensators:  
one or more closed-loop poles near zero - sluggish response!

9a



rlocus (n, d)  
rlocfind (n, d)

96

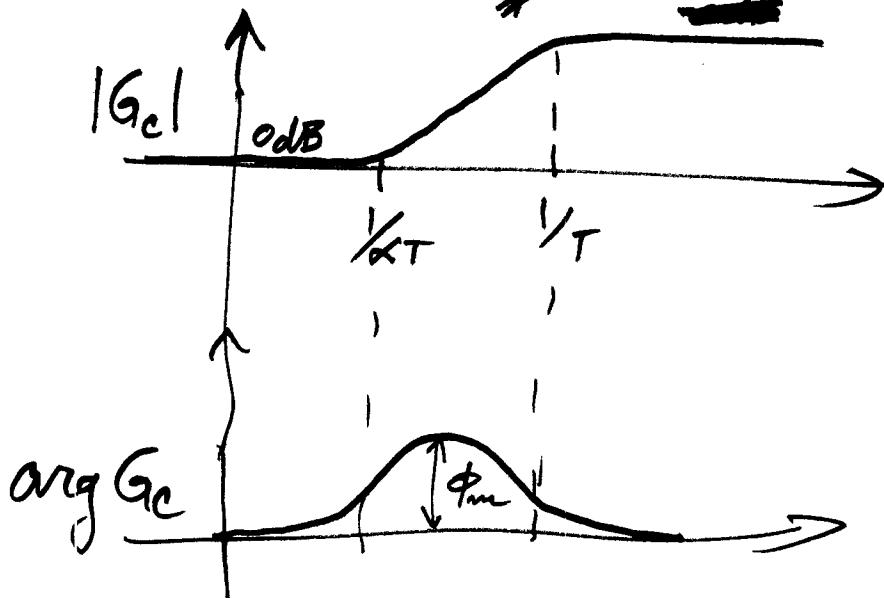


$$[nc, dc] = \text{cloop}(n, d, -1)$$

Step(nc, dc)

# Phase-Lead Compensator

$$G_C(s) = \frac{\alpha T s + 1}{T s + 1}, \quad \alpha > 1$$



Back to example on p. 7a

$$G(s) = \frac{100}{s(s+4)(s+5)}$$

The max. phase lead  $\phi_m$  should be achieved at the new crossover frequency  $\bar{\omega}_c$ .   
 $\bar{\omega}_c$  moves to the right because of the gain increase. Pick  $\alpha$  and  $T$  so that  $\omega_c$  is between  $\frac{1}{\alpha T}$  and  $\frac{1}{T}$ , closer to  $\frac{1}{\alpha T}$ .

Since  $\omega_c = 3.25$ , we select

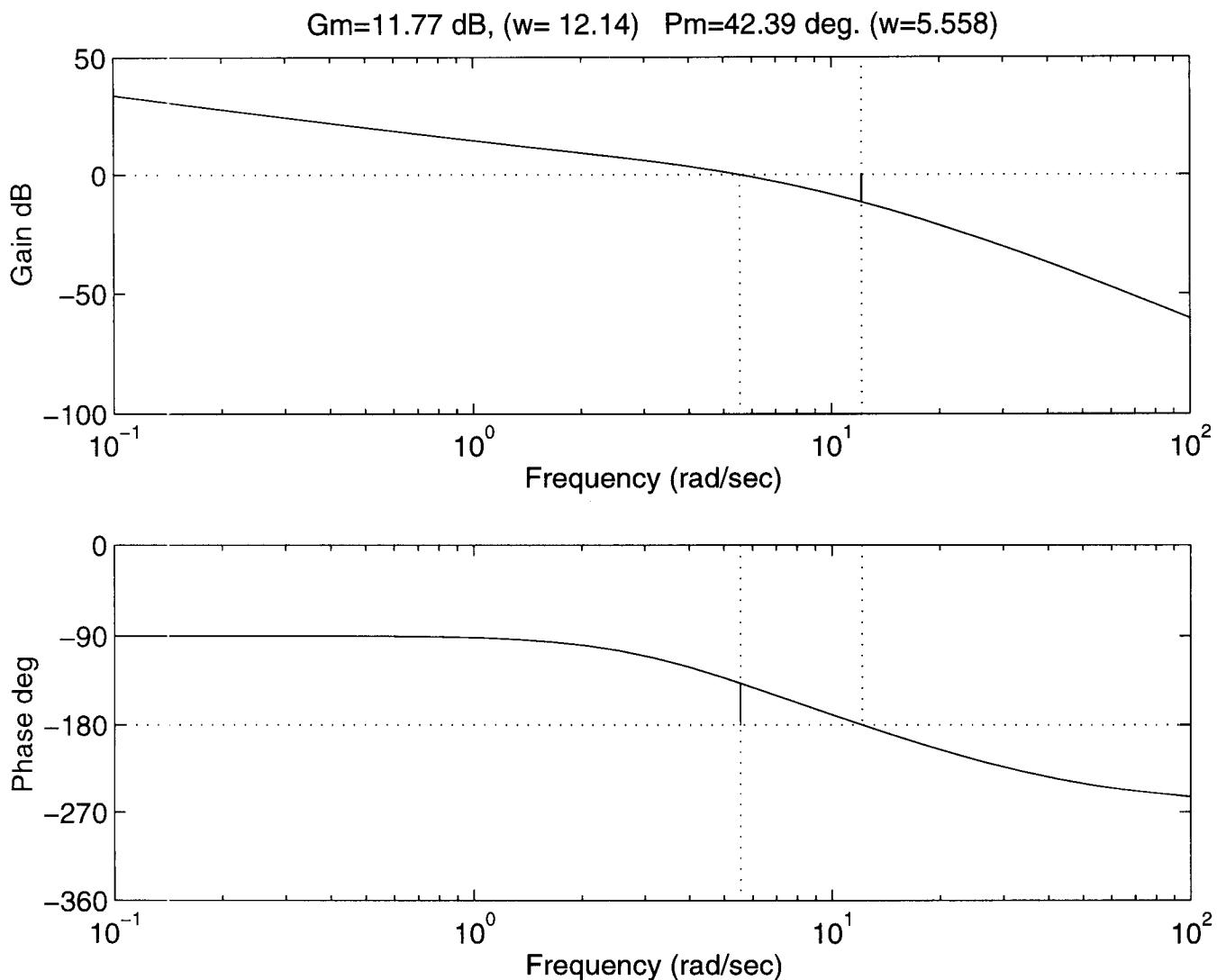
$$\begin{aligned} G_C(s) &= \frac{s+2}{s+20} \frac{20}{2} \\ &= \frac{0.5s + 1}{0.05s + 1} \end{aligned}$$

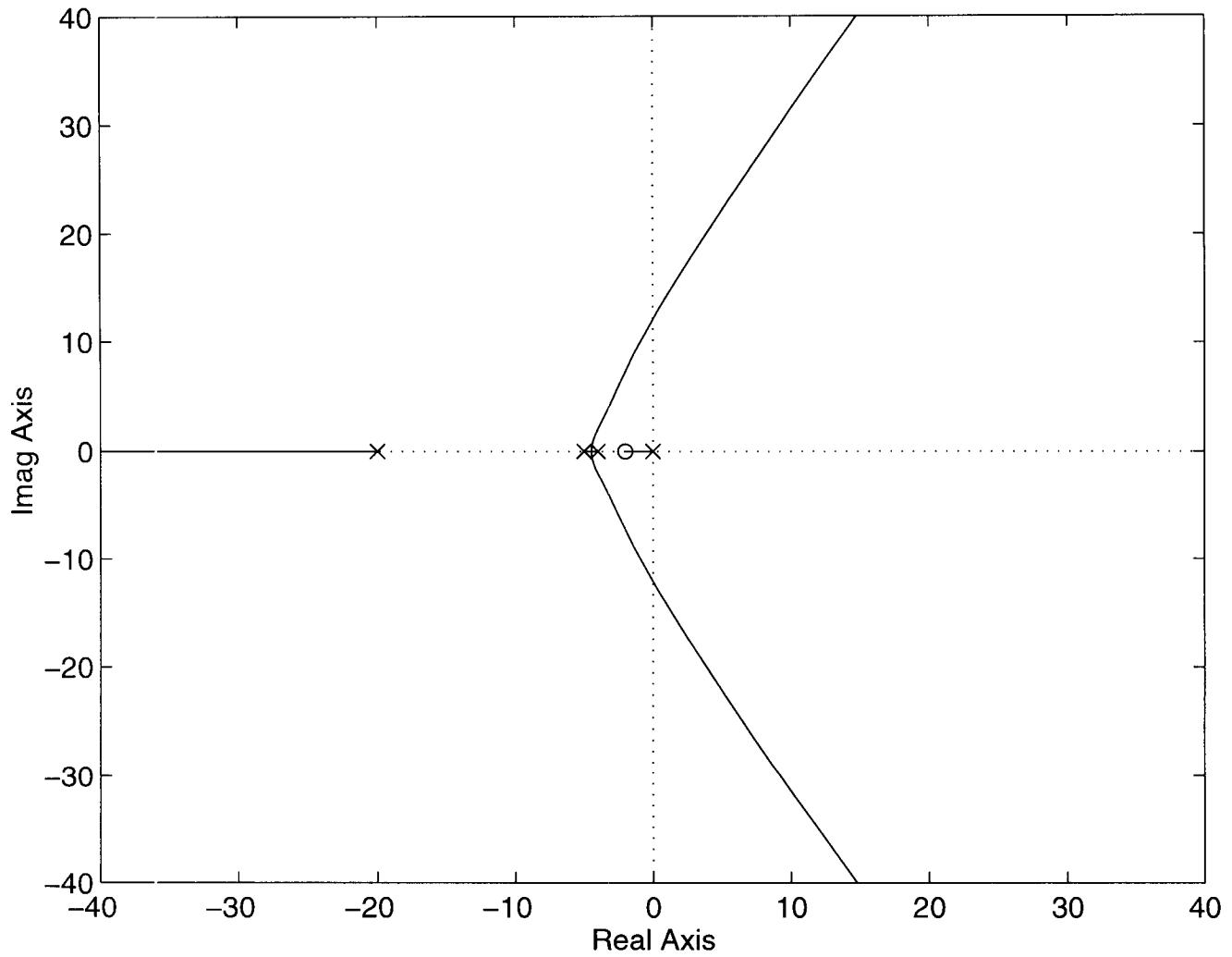
The new  $\bar{\omega}_c = 5.5$  and  $PM = 42.4$

The rise time is about 5 times faster!

11a

$$G_C(s) G(s) = \frac{0.5s + 1}{0.05s + 1} \frac{100}{s(s+4)(s+5)}$$





11c

