## **Robust Adaptive Control** and the **Greeks** who Made it Possible

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Advances in Design and Control

### Adaptive Backstepping

Nonlinear and Adaptive Control Design

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imon Haykin, Series Editor

- Nonlinear & Adaptive Control Design—1995
- adapt. backstepping & strict-fbk systems (1991): Kanellakopoulos, Kokotovic, Morse
- prescribed performance (2008): Bechlioulis & Rovithakis

## **Robust Adaptive Control:**

40+ years since the KEY (unimproved) ideas

### **Robust Adaptive Control** — the Ordeal

- adaptive control stability (MRAC) solved  $\approx$  1980
- Rohrs-Athans (1982) unstable simulations (disturbances, unmod. dynamics)
- deadzone: Peterson-Narendra-Annaswamy (1982-1989)

small disturbance, requires PE

parameter projection

must know parameter bound

• *σ*-modification:

Ioannou (1983–1996), + Kokotovic, Tsakalis/Tao/Sun/Datta/Polycarpou, etc.

last 30 years of my editorial observation of adaptive (nonlinear) control:
 σ-modification owns 95% of the "robustification market"

## "IOANNOU effect"

## Effect of 1996 book by loannou & Sun

- Systematizes, comprehensively covers, and advances linear (robust) adaptive control to date (800 pages)
- · Deals not only with disturbances but also (multiplicative) unmodeled dynamics,

$$y = G_0(s) \left(1 + \Delta_m(s)\right) u \,.$$

Example of a theorem in [IS'96], roughly expressed: **Theorem 9.3.3.** If

$$\min\left\{1 + \frac{\|\Delta_m\|_{\infty\delta_0}^2}{\delta_0^{2(n^*+1)}}, \sigma + \|H\Delta_m\|_{2\delta_0}^2\right\} \quad \text{suffic. small}$$

then the tracking error e(t) is bounded-in-the-mean, i.e.,

$$\frac{1}{T}\int_{t}^{t+T}e^{2}(\tau)d\tau \leq c\left(\frac{1}{T}+\sigma+\|\Delta_{m}\|_{\infty\delta_{0}}^{2}+\|H\Delta_{m}\|_{2\delta_{0}}^{2}\right)$$

Results dishearteningly clear and precise regarding

- the complexity of the robust adaptive problem,
- the technical virtuosity required to keep making progress.

Disturbance-Robustness under  $\sigma$ -Modification

$$\dot{x} = \theta x + u + d$$

Adaptive controller with leakage:

$$u = -cx - \hat{\theta}x$$
$$\dot{\hat{\theta}} = \Gamma x^2 - \sigma \hat{\theta}$$

## p-IOS and p-OAG

Lyapunov calculation yields practical input-output-stability (p-IOS) w.r.t. disturbance d:

$$|x(t)| \leq e^{-\sigma t/2} \left( |x_0| + \left| \hat{\theta}_0 - \theta \right| \right) + \frac{\sqrt{\Gamma}}{\sigma} ||d||_{\infty} + |\theta|$$

Additionally, practical output asymptotic gain (p-OAG):

$$\limsup_{t \to +\infty} |x(t)| \le \frac{\sqrt{\Gamma}}{\sigma} \limsup_{t \to +\infty} |d(t)| + |\theta|$$

The p-OAG  $\sqrt{\Gamma/\sigma}$  can be reduced (though not to zero) by increasing leakage  $\sigma$ . But the bias  $|\theta|$  is neither known nor reducible.

## The culprit for the bias $|\theta|$ is the leakage itself

• The term  $|\Gamma x^2|$  in the update

$$\dot{\hat{\theta}} = \boxed{\Gamma x^2} - \sigma \hat{\theta}$$

pumps up  $\hat{\theta}$ , i.e., pumps error  $\tilde{\theta} = \theta - \hat{\theta}$  to *negative*/disturbance-attenuating values.

• But in the parameter error system

$$\dot{\tilde{\theta}} = -\sigma\tilde{\theta} - \Gamma x^2 + \sigma\theta$$

the term  $+\sigma\theta$  counters the work of  $\Gamma x^2$ .

 $\sigma$ -modification's task CONTRADICTS the task of adaptation.



 $\begin{array}{l} \mbox{Fig. 1. } (y, \hat{\theta}, u) \mbox{ for closed-loop system with } \pmb{\sigma}\mbox{-mod}, \\ & \mbox{ with } d(t) = 2 \sin(\pi t), \ \theta = 3, \\ & \ c = 1, \ \Gamma = 10, \ \sigma = q = 0.5, \\ & \ y_0 = 0.5, \ \hat{\theta}(0) = 0. \end{array}$ 

## **DADS: the Scalar Case**

(DADS = Deadzone-Adapted Disturbance Suppression)

### The gist of what you will see:

$$\dot{x} = \theta x + u + d$$

$$u = -x - \kappa (2 + x^2) x$$

$$\dot{\kappa} = \Gamma \max \{0, x^2 - \delta^2\}$$

guarantees

$$\limsup_{t \to +\infty} |x(t)| \le \delta$$

for arbitrarily small  $\delta > 0$  independent of  $\sup_{t>0} |d(t)|$  and  $|\theta|$ ,



Fig. 1.  $(y, \hat{\theta}, u)$  for closed-loop system with  $\sigma$ -mod, with  $d(t) = 2\sin(\pi t), \theta = 3$ ,  $c = 1, \Gamma = 10, \sigma = q = 0.5$ ,  $y_0 = 0.5, \hat{\theta}(0) = 0$ . Fig. 2.  $(y, \hat{\theta}, u)$  for DADS closed-loop system, with  $d = 2 \sin(\pi t)$ ,  $\theta = 3$ , c = 1/2, q = 1/4,  $\Gamma = 50$ ,  $\delta = 0.01$ ,  $y_0 = 0.5$ ,  $z_0 = -\ln(10)$ .

**TENFOLD** improved disturbance suppression under DADS, using the SAME CONTROL EFFORT, relative to  $\sigma$ -modification in Figure 1.

- <u>update</u> "pumps up" the controller dynamic gain  $\kappa$  to a sufficient size, adapting the gain to the unknown disturbance d,
- deadzone shuts off further pumping up of the gain once the state is in the prespecified residual set.

Unlike  $\sigma$ -mod, the <u>deadzone</u>'s task <u>does not contradict</u> the <u>adaptation</u>'s task.

In design:

- deadzone is not new
- new uses of completing squares (dynamic nonlinear damping, to obliterate ("practically," ultimately) the effects of disturbance and parametric uncertainty

In analysis:

• A new comparison lemma (I call it "Karafyllis lemma")

## Notation for deadzone function $(\cdot)^+$

$$s^+ := \max\{0, s\}$$
 = ReLU(s)

deadzone acting on

 $s = x^2 - \delta^2$ 

ReLU — notation common in neural networks, representing the "rectifier linear unit"

## **Control** Design Idea

### Recall the adaptive controller



## What drives this Control Law Design?

Lyapunov function (independent of gain  $\kappa$ )

$$V(x) = \frac{x^2}{2}$$

Control to employ dynamic nonlinear damping and ensure

$$\dot{V} \leq -2cV + \frac{d^2 + \left( \left( |\theta| - \kappa \right)^+ \right)^2}{4q\kappa}$$

This form of inequality ensures

- ISS w.r.t. d
- gain  $\kappa(t)$  ultimately dominates the unknown  $|\theta|$ ; makes  $x(t) \rightarrow 0$  when  $d(t) \rightarrow 0$

Stability/Convergence Properties

### **Basic results from Lyapunov inequality**

Practical IOS  
$$|x(t)| \le e^{-ct} |x_0| + \frac{1}{2\sqrt{cq}} ||d||_{\infty} + \frac{1}{2\sqrt{cq}} \left( \frac{|\theta|}{|\theta|} - 1 \right)^+$$

(Practical) Asymptotic Gain to Disturbance

$$\limsup_{t \to +\infty} |x(t)| \le \frac{1}{2\sqrt{cq}} \left(\limsup_{t \to +\infty} |d(t)| + (|\theta| - \kappa_{\infty})^{+}\right)$$

#### Theorem

The DADS closed-loop system satisfies



assignable gain to d, with bias dependent on  $|\theta|$ 

All of these also achieved with  $\sigma$ -mod.

## No Parameter Drift

## **No Parameter Drift**

This is the most novel and complex part of our analysis approach.

Lemma (Comparison lemma on deadzone in "exterminator-pest" feedback)

Consider the differential inequalities

$$0 \leq \dot{\kappa} \leq (V - \varepsilon)^{+}, \quad \kappa_{0} > 1$$
  
$$\dot{V} \leq -V + \frac{\mu}{\kappa}, \quad V_{0} \geq 0$$

Then,

$$\kappa_0 \leq \kappa(t) \leq \max\left\{\kappa_0, \frac{\mu}{\varepsilon}\right\} + \left(V_0 - \varepsilon + \frac{\mu}{\kappa_0}\right)^+, \quad \forall t \geq 0.$$
 Karafyllis

**Pest** (V) grows then decays.

**Exterminator** ( $\kappa$ ) grows but bounded.

## Explicit bound on parameter update

Practical uniform bounded-input bounded state (p-UBIBS) property:

$$\kappa_0 \leq \kappa(t) \leq \kappa_{\infty} \leq \max\left\{\kappa_0, \frac{\mu}{4\delta^2}\right\} + \frac{1}{2}\left(x_0^2 - \delta^2 + \frac{\mu}{4\kappa_0}\right)^+, \quad \forall t \geq 0$$
  
where  
$$\mu = ||d||_{\infty}^2 + \left((|\theta| - \kappa_0)^+\right)^2.$$
The bound on  $\kappa(t)$  is  
• increasing in  $||d||_{\infty}$  and  $|x_0|$   
• non-decreasing in  $|\theta|$   
and  
• increasing in  $1/\delta$ 

(The tighter the asymptotic regulation  $\delta$  desired, the higher  $\kappa(t)$  needs to be pumped.)

#### Theorem

The DADS closed-loop system satisfies

p-UBIBS	no parameter drift		
p-IOS	assignable gain to a	l, with	bias dependent on $ \theta $



All of these also achieved with  $\sigma$ -mod.

## Zero Asymptotic Gain to Disturbance

## **Practical Regulation for Arbitrary Disturbance**

# By the Karafyllis lemma - and - Barbalat's lemma for $\dot{\kappa}(t) = \Gamma (x^2(t) - \delta^2)^+$ , we obtain the zero practical asymptotic gain (zero p-OAG) property

$$\limsup_{t \to +\infty} |x(t)| \le \frac{\delta}{\delta}$$

## **Perfect Regulation**

## under Asymptotically Vanishing Disturbance

When  $\lim_{t\to+\infty} d(t) = 0$  (the disturbance-vanishing case),

### $\sigma$ prevents convergence of x(t) to 0

periect regulation is achieved,

$$\lim_{t\to+\infty} x(t) = 0$$

To increase the *chance* of  $\kappa_{\infty} \geq |\theta|$ , increase  $\Gamma$ 

## Perfect Regulation under Asymptotically Vanishing Disturbance

When  $\lim_{t\to+\infty} d(t) = 0$  (the disturbance-vanishing case),

and when either  $|\theta| \leq 1$  or  $\kappa_{\infty} \geq |\theta|$ ,

(similar to switching-σ)

perfect regulation is achieved,

$$\lim_{t \to +\infty} x(t) = 0$$

To increase the *chance* of  $\kappa_{\infty} \geq |\theta|$ , increase  $\Gamma$ .

Summary: MAIN RESULT

#### Theorem

#### The DADS closed-loop system satisfies

p-UBIBS	no parameter drift				
p-IOS	assignable gain to $d$ , with bias dependent on $ \theta $				
zero p-OAG	residual error indep. of $(d, \theta)$ and assignably small				
$x(t) \rightarrow 0$	when $d(t) \rightarrow 0$ and $\kappa_{\infty}$ large				
IOS w/ small $ \theta $	IOS gain & asymptotic gain to d are assignable				

#### The properties in BLUE are NOT guaranteed with $\sigma$ -mod.







## BACKSTEPPING for Strict-Feedback Systems

## DADS Backstepping for Strict-Feedback Systems 🔸

#### Theorem

Consider the strict-feedback system

**UNMATCHED** parametric uncertainty

$$\dot{x}_i = x_{i+1} + \varphi'_i(x_1, ..., x_i)\theta + \alpha'_i(x_1, ..., x_i)d$$
  $i = 1, ..., n, \quad x_{i+1} = u$ 

Explicit backstepping design guarantees, globally, the p-OAG, p-IOS, p-UBIBS properties

$$\begin{split} \limsup_{t \to +\infty} |x_1(t)| &\leq \varepsilon \\ |x_1(t)|^2 &\leq 2e^{-2c_0 t} V_n(x_0, z_0) + \frac{\|d\|_{\infty}^2 + \left( (\|\theta\|_{\infty} - 1 - e^{z_0})^+ \right)^2}{4c_0 \beta_0 (1 + e^{z_0})} \\ |x(t)| &\leq B(x_0, z_0, \|d\|_{\infty}, \|\theta\|_{\infty}) \\ z_0 &\leq z(t) &\leq \lim_{s \to +\infty} z(s) \leq B(x_0, z_0, \|d\|_{\infty}, \|\theta\|_{\infty}) \end{split}$$

## **Explicit DADS Backstepping**

#### Limit cycle at high angle-of-attack

roll angle	$\dot{x}_1 = x_2$
roll rate	$\dot{x}_2 = \varphi^{\mathrm{T}}(x_1, x_2)\theta + x_3 + d_2$
aileron angle	$\dot{x}_3 = u + d_3$

 $\varphi(\text{roll-angle, roll-rate}) = \text{aerodynamic forces vector}$ 

disturbance vector d(t) = taken sinusoidal for simulations

## Aircraft Wing Rock



### But what about robustness to NOISE?

plant:	x	=	$\theta x + u$
output:	y	=	x + n(t)
controller:	и	=	$-y-\kappa\left(2+y^2\right)y$
gain update:	Ķ	=	$\Gamma \left( \boldsymbol{y^2} - \delta^2 \right)^+$

$$\limsup_{t \to +\infty} |x(t)| \le \frac{\delta}{t} + \limsup_{t \to +\infty} |n(t)|$$

$$\kappa_0 \le \kappa(t) \le \kappa_\infty \le \max\left\{\kappa_0, \frac{\mu}{4\delta^2}\right\} + \frac{1}{2}\left(\|\boldsymbol{n}\|_{\infty}^2 - \delta^2 + \frac{\mu}{4\kappa_0}\right)^+$$
$$\mu = \left\|\frac{\mathrm{d}\boldsymbol{n}}{\mathrm{d}t}\right\|_{\infty}^2 + |\boldsymbol{\theta}|\|\boldsymbol{n}\|_{\infty}^2 + \left((|\boldsymbol{\theta}| - \kappa_0)^+\right)^2$$

- tracking treats reference as a disturbance (brute force), due to state reference trajectory being affected by unknown parameters
- output feedback adaptive observer error likely to manifest itself as a disturbance and be tractable
- unmodeled dynamics, ISS/small-gain

Adaptive control can achieve arbitrarily good regulation

- for arbitrarily large disturbances
- without PE

No "fundamental limitation," NO "CURSE OF UNCERTAINTY"!